

24V, 150mA, Low Quiescent Current Linear Regulator

DESCRIPTION

The MP2011A is a low power linear regulator that supplies power to systems with high voltage batteries. It includes a wide 2.5V to 24V input range, low dropout voltage and low quiescent supply current. The low quiescent current and low dropout voltage allow operations at extremely low power levels. Therefore, the MP2011A is an ideal for the low power microcontrollers and the battery-powered equipments.

The MP2011A provides wide variety of fixed output voltage options (if request): 1.8V, 1.9V, 2.3V, 2.5V, 3.0V, 3.3V, 3.45V, 5.0V; and also provides the output adjustable option (from 1.215V to 15V).

The regulator output current is internally limited and the device is protected against over-load and over-temperature conditions.

The MP2011A also includes thermal shutdown and current limiting fault protection, and is available in QFN-6 (2mmx2mm) packages.

FEATURES

- 2.5V to 24V Input Range
- 3.3µA Quiescent Supply Current
- Stable With low-value output ceramic capacitor (> 0.47μF)
- 150mA load current
- 700mV Dropout at 150mA Load
- Available in Fixed 3.3V Output and Adjustable Output (1.215V to 15V) versions
- Output ±1.5% Accuracy
- Specified current limit
- Thermal Shutdown and Short-Circuit Protection
- Available in QFN-6 (2mmx2mm) Packages

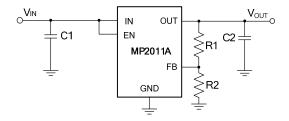
APPLICATIONS

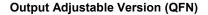
- Portable/Battery-Powered Equipment
- Ultra low power Microcontrollers
- Cellular Handsets

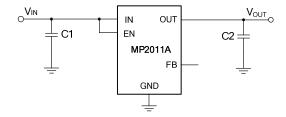
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TYPICAL APPLICATION







Output Fixed Version (QFN)



ORDERING INFORMATION

Part Number*	Package	Top Marking
MP2011AGG	QFN-6 (2mmx2mm)	See Below
MP2011AGG-33	QFN-6 (2mmx2mm)	See Below

^{*} For Tape & Reel, add suffix -Z (e.g. MP2011AGG-Z);

TOP MARKING (MP2011AGG)

DDY

LLL

DD: product code of MP2011AGG;

Y: year code; LLL: lot number;

TOP MARKING (MP2011AGG-33)

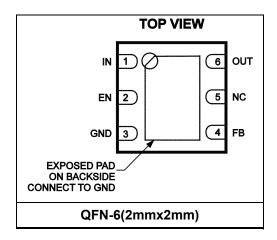
DCY

LLL

DC: product code of MP2011AGG-33;

Y: year code; LLL: lot number;

PACKAGE REFERENCE



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ABSOLUTE MAXIMUM RATINGS (1)
IN, EN0.3V to +26V
OUT0.3V to +17V
FB0.3V to +6V
Lead Temperature260°C
Storage Temperature65°C to +150°C
Continuous Power Dissipation $(T_A = +25^{\circ}C)^{(2)}$
QFN-6(2mmx2mm)1.5W
ESD SUSCEPTIBILITY (3)
HBM (Human Body Mode)2kV
MM (Machine Mode)200V
Recommended Operating Conditions (4)
Supply Voltage V _{IN} 2.5V to 24V
Output Voltage V _{OUT} 1.215V to 15V
Operating Junction Temp (T _J)40°C to +125°C

Thermal Resistance ⁽⁵⁾	$oldsymbol{ heta}_{JA}$	$\boldsymbol{\theta}_{JC}$	
QFN-6 (2mmx2mm)	80	16	.°C/W

Notes:

- 1) Exceeding these ratings may damage the device.
- 2) The maximum allowable power dissipation is a function of the maximum junction temperature $T_{\rm J}$ (MAX), the junction-to-ambient thermal resistance $\theta_{\rm JA}$, and the ambient temperature $T_{\rm A}$. The maximum allowable continuous power dissipation at any ambient temperature is calculated by $P_{\rm D}$ (MAX) = ($T_{\rm J}$ (MAX)- $T_{\rm A}$)/ $\theta_{\rm JA}$. Exceeding the maximum allowable power dissipation will cause excessive die temperature, and the regulator will go into thermal shutdown. Internal thermal shutdown circuitry protects the device from permanent damage.
- Devices are ESD sensitive. Handling precaution recommended.
- The device is not guaranteed to function outside of its operating conditions.
- 5) Measured on JESD51-7, 4-layer PCB.



ELECTRICAL CHARACTERISTICS

 $V_{EN}=V_{IN}$, $C_{OUT}=1\mu F$, $T_J=-40^{\circ}C$ to +125°C, Typical value is tested at $T_J=25^{\circ}C$. The limit over temperature is guaranteed by characterization, unless otherwise noted.

Parameter	Symbol	Condition	Min	Тур	Max	Units
Input Voltage	V_{IN}		2.5		24	V
Output Voltage Range	V _{OUT}		1.215		15	V
GND Pin Current		MP2011AGG, $0 < I_{OUT} < 1mA$, $V_{IN} = 2.5V$ to 24V, $V_{OUT} = 5V$ ($V_{IN} \ge 6V$) or FB ($V_{IN} < 6V$)		3.3	9	
		MP2011AGG, 1mA< I_{OUT} <30mA, V_{IN} =2.5V to 15V, V_{OUT} = 5V (V_{IN} >6V) or FB (V_{IN} <6V)		11	22	uA
	I _{GND}	MP2011AGG, 30mA <i<sub>OUT<150mA, V_{IN}=3V, V_{OUT}=FB</i<sub>		32	50	
		MP2011AGG-33, 0mA <i<sub>OUT<1mA, V_{IN}=4.3V to 24V</i<sub>		4.4	10	
		MP2011AGG-33, 1mA <i<sub>OUT<30mA, V_{IN}=4.3V to 15V</i<sub>		15	25	uA
		MP2011AGG-33, 30mA <i<sub>OUT<150mA , V_{IN}=4.3V</i<sub>		35	50	
Shutdown Supply Current	I _{SHDN}	V _{EN} =0,V _{IN} =2.5V to 24V		3	9	uA
Load Current		MP2011AGG-33, V _{OUT} =0V, V _{IN} =4.3V to 15V	160	270	400	mA
limit	I _{LIMIT}	MP2011AGG, V _{OUT} =0V, V _{IN} =6V to 15V	160	270	400	mA
Output Voltage Accuracy		MP2011AGG-33, V _{IN} =4.3V, I _{OUT} =0, T _J =25°C	3.251	3.3	3.349	
		MP2011AGG- 33,V _{IN} =4.3V,I _{OUT} =0,T _J =-40°C to 125°C	3.168	3.3	3.432	V
FB Voltage	V _{FB}	MP2011AGG/AGG-33, FB=OUT, V_{IN} =6V, I_{OUT} =0, T_{J} =25°C	1.197	1.215	1.233	V
		MP2011AGG/AGG-33, FB=OUT, V_{IN} =6V, I_{OUT} =0, T_{J} =-40°C to 125°C	1.167	1.215	1.263	V
Dropout Voltage		MP2011AGG, I _{OUT} =150mA, V _{OUT(NOM)} =5V		600	1300	mV
V _{IN} =V _{OUT(NOM)} - 0.1V	V _{DROPOUT}	MP2011AGG-33, I _{OUT} =150mA, V _{OUT(NOM)} =3.3V		700	1300	mV



ELECTRICAL CHARACTERISTICS (continued)

 $V_{EN}=V_{IN}$, $C_{OUT}=1\mu F$, $T_J=-40^{\circ}C$ to +125°C, Typical value is tested at $T_J=25^{\circ}C$. The limit over temperature is guaranteed by characterization, unless otherwise noted.

I _{FB}	MP2011AGG-33, V_{FB} =1.3V, V_{IN} =6V, OUT floating. MP2011AGG, V_{FB} =1.3V, V_{IN} =6V, OUT floating. V_{IN} =2.5V to 24V, I_{OUT} =1mA, OUT=FB	0.88 -50	1.3	1.72 50	uA
'I-В	V _{IN} =6V, OUT floating. V _{IN} =2.5V to 24V, I _{OUT} =1mA, OUT=FB	-50	0	50	
	OUT=FB				nA
			0.01	0.05	%/V
	MP2011AGG, I _{OUT} =100uA to 150mA, V _{IN} =3V to 6V		0.005	0.012	%/mA
	MP2011AGG-33, I _{OUT} =100uA to 150mA, V _{IN} =4.3V to 6V		0.005	0.020	%/mA
	100Hz, C_{IN} =100pF, C_{OUT} =4.7 μ F, I_{OUT} =10mA, V_{IN} =6V, V_{OUT} =5V		58		dB
	1kHz, C _{IN} =100pF, C _{OUT} =4.7μF, I _{OUT} =10mA, V _{IN} =6V, V _{OUT} =5V		41		dB
	100kHz, C _{IN} =100pF, C _{OUT} =4.7μF, I _{OUT} =10mA, V _{IN} =6V, V _{OUT} =5V		55		dB
	MP2011AGG, I _{OUT} =100mA, C _{OUT} =6.8uF, V _{OUT} =5V			3	ms
	MP2011AGG-33, I _{OUT} =10mA, C _{OUT} =6.8uF, V _{OUT} =3.3V			1.8	ms
EN_{TH_R}	V _{IN} =2.5V to 24V	1.32	1.48	1.64	V
EN _{TH_F}	V _{IN} =2.5V to 24V	1.1	1.26	1.42	V
IEN	V _{EN} =0 or 15V, V _{IN} =2.5V to 24V			0.1	uA
TSD		150	165		С
∆TSD			20		С
	EN _{TH_F} IEN TSD	I _{OUT} =100uA to 150mA, V _{IN} =4.3V to 6V 100Hz, C _{IN} =100pF, C _{OUT} =4.7μF, I _{OUT} =10mA, V _{IN} =6V, V _{OUT} =5V 1kHz, C _{IN} =100pF, C _{OUT} =4.7μF, I _{OUT} =10mA, V _{IN} =6V, V _{OUT} =5V 100kHz, C _{IN} =100pF, C _{OUT} =4.7μF, I _{OUT} =10mA, V _{IN} =6V, V _{OUT} =5V MP2011AGG, I _{OUT} =100mA, C _{OUT} =6.8uF, V _{OUT} =5V MP2011AGG-33, I _{OUT} =10mA, C _{OUT} =6.8uF, V _{OUT} =3.3V EN _{TH_R} V _{IN} =2.5V to 24V IEN V _{EN} =0 or 15V, V _{IN} =2.5V to 24V TSD	I _{OUT} =100uA to 150mA,	$\begin{array}{c ccccccccccccccccccccccccccccccccccc$	I _{OUT} =100uA to 150mA,

Notes:

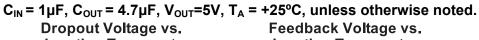
6) Line Regulation=
$$\frac{|V_{\text{OUT}[V_{\text{IN}(\text{MAX})}]} - V_{\text{OUT}[V_{\text{IN}(\text{MIN})}]}|}{(V_{\text{IN}(\text{MAX})} - V_{\text{IN}(\text{MIN})}) \times V_{\text{OUT}(\text{NOM})}} \times (\% / V)$$

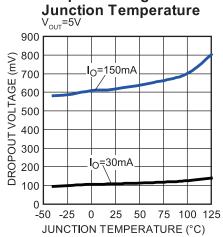
7) Load Regulation=
$$\frac{|V_{\text{OUT}[l_{\text{OUT}(\text{MAX})}]} - V_{\text{OUT}[l_{\text{OUT}(\text{MIN})}]}|}{(I_{\text{OUT}(\text{MAX})} - I_{\text{OUT}(\text{MIN})}) \times V_{\text{OUT}(\text{NOM})}} \times (\%/\text{mA})$$

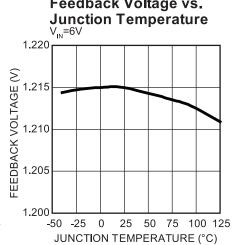
8) Derived from bench characterization. Not tested in production.

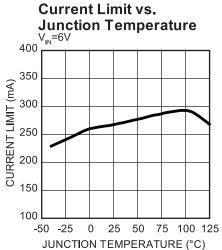


TYPICAL CHARACTERISTICS





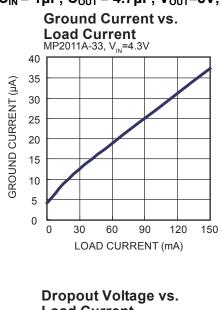


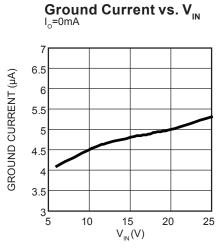


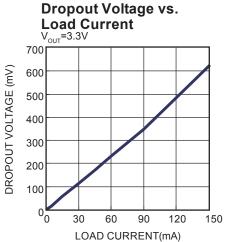


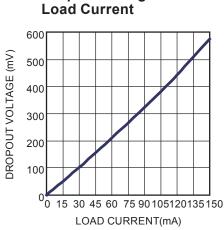
TYPICAL PERFORMANCE CHARACTERISTICS

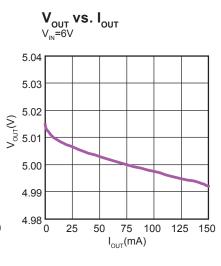
 C_{IN} = 1 μ F, C_{OUT} = 4.7 μ F, V_{OUT} =5V, T_A = +25°C, unless otherwise noted

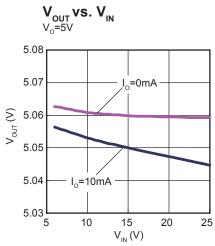


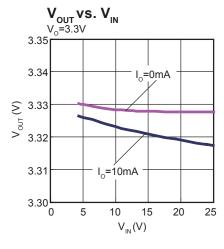


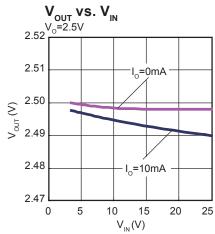


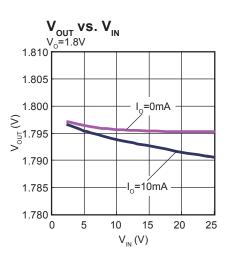








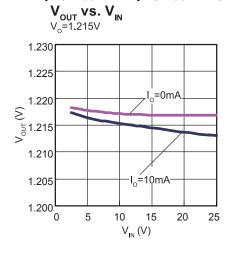


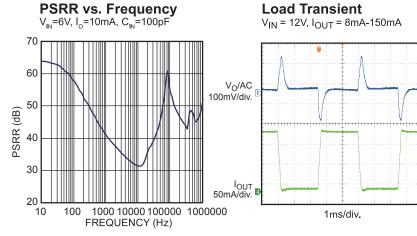


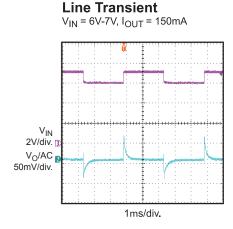


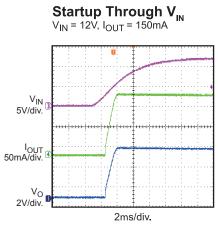
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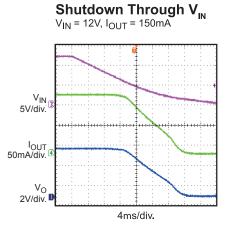


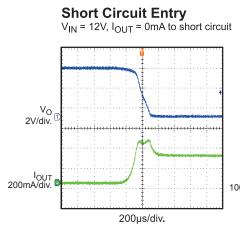


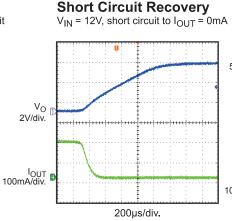


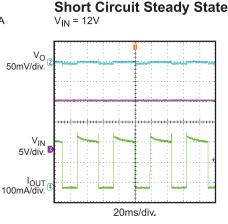














PIN FUNCTIONS

Pin#		
QFN6 (2x2mm)	Name	Description
1	IN	Input Voltage. Connect a 2.5V to 24V supply to this pin.
2	EN	A logic low on this pin will shut down the IC, a logic high will start it up. Connect this pin to IN for automatic start up.
3, exposed pad	GND	Ground (the exposed pad and GND pin must be connected to the same ground plane)
4	FB	Feedback Input Pin, regulated to 1.215V nominally. Connected to an external resistive divider between OUT and GND to set output voltage. For fixed output version, this pin can be just float.
5	NC	No Connection. May be left open or tied to Ground for improved thermal performance.
6	OUT	Regulated Output Voltage, only low-value ceramic capacitor (≥ 0.47µF) on output is required for stability.



OPERATION

The MP2011A is a linear regulator that supplies power to systems with high voltage batteries. It includes a wide 2.5V to 24V input range, low dropout voltage and low quiescent supply current.

The MP2011A provides wide variety of fixed output voltage option (if request): 1.8V, 1.9V, 2.3V, 2.5V, 3.0V, 3.3V, 3.45V, 5.0V; and also provides the output adjustable option (from 1.215V to 15V).

The output adjustable version has an output that is adjustable from 1.215V to 15V with a simple resistor divider. It uses external feedback, allowing the user to set the output voltage with an external resistor divider. The typical FB pin voltage is 1.215V.

The regulator output current is internally limited and the device is protected against over-load and over-temperature conditions.

The peak output current is limited to around 270mA, which exceeds the 150mA recommended continuous output current.

When the junction temperature is too high, the thermal sensor sends a signal to the control logic that will shutdown the IC. The IC will restart when the temperature has sufficiently cooled.

The maximum power output current is a function of the package's maximum power dissipation for a given temperature.

The maximum power dissipation is dependent on the thermal resistance of the case and the circuit board, the temperature difference between the die junction and the ambient air, and the rate of air flow. The GND pin and Exposed Pad must be connected to the ground plane for proper dissipation.

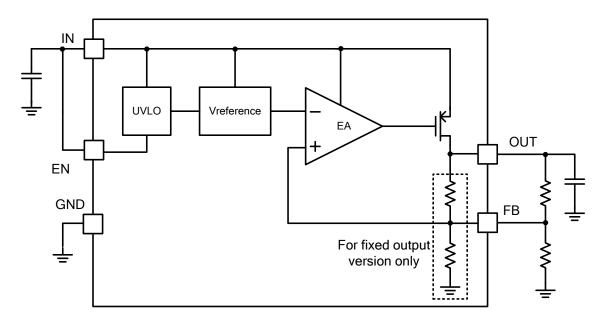


Figure 1: Functional Block Diagram



APPLICATION INFORMATION

COMPONENT SELECTION

Setting the Output Voltage

For adjustable output version. Set the output voltage of the MP2011A by using a resistor divider as shown:

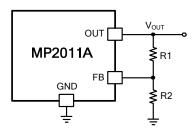


Figure 2: FB Resistor Divider to Set Vout

Choose R2=1M Ω to maintain a 1.215 μ A minimum load. Calculate the value for R1 using the following equation:

R1=R2×
$$\left(\frac{V_{OUT}}{1.215V}-1\right)$$
 (1)

The above equation is only valid on adjustable output version. For fixed output version, V_{OUT} also can be adjusted by adding external resistor divider. Just note to take internal FB resistor divider into consideration when choose external divider. Choose a small resistance to mask internal feedback resistor network. The equation (1) still work but require the sum of R1 and R2 is less than $100 \text{k}\Omega$.

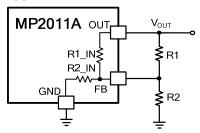


Figure 3: FB Divider of Fixed Output Version

The internal FB resistor dividers for different fixed output versions please see below table.

Table 1: Internal FB Resistor Divider

Fixed Output Voltage	R1_IN	R2_IN
3.3V	1.72ΜΩ	1ΜΩ

Input Capacitor

For proper operation, place a ceramic capacitor (C1) between $1\mu F$ and $10\mu F$ of dielectric type X5R or X7R between the input pin and ground. Larger values in this range will help to improve line transient response.

Output Capacitor

For stable operation, use a ceramic capacitor (C2) of type X5R or X7R between $1\mu F$ and $10\mu F$. Larger values in this range will help to improve load transient response and reduce noise. Output capacitors of other dielectric types may be used, but are not recommended as their capacitance can deviate greatly from their rated value over temperature.

To improve load transient response, add a small ceramic (X5R, X7R or Y5V dielectric) 22nF feed forward capacitor in parallel with R1. The feed forward capacitor is not required for stable operation.

Output Noise

The MP2011A will exhibit noise on the output during normal operation. This noise is negligible for most applications. However, in applications that include analog-to-digital converters (ADCs) of more than 12 bits, one needs to consider the ADC's power supply rejection specifications. The feed forward capacitor across R1 will significantly reduce the output noise.

PCB layout guide

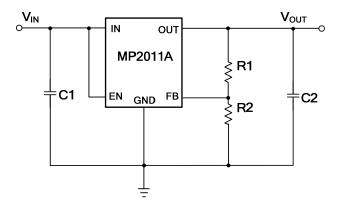
PCB layout is very important to achieve good regulation, ripple rejection, transient response and thermal performance. It is highly recommended to duplicate EVB layout for optimum performance.

If change is necessary, please follow these guidelines and take figure 4 for reference.

- Input and output bypass ceramic capacitors are suggested to be put close to the IN Pin and OUT Pin respectively.
- Ensure all feedback connections are short and direct. Place the feedback resistors and compensation components as close to the chip as possible.



 Connect IN, OUT and especially GND respectively to a large copper area to cool the chip to improve thermal performance and long-term reliability.



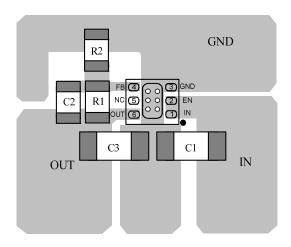


Figure 4: PCB Layout (Top Layer)

Design Example

Below is a design example following the application guidelines for V_{OUT} =3.3V.

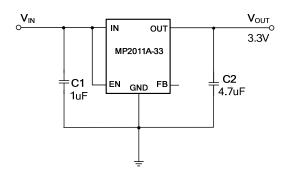


Figure 5: Design Example



TYPICAL APPLICATION CIRCUITS

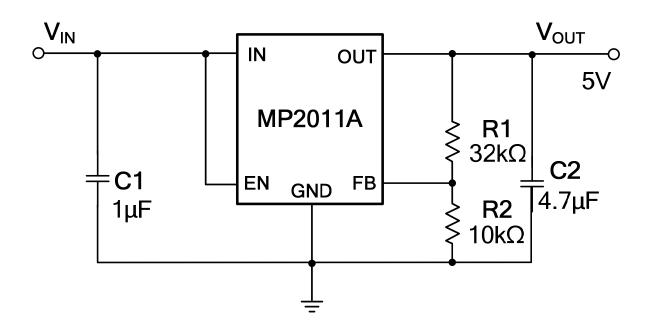
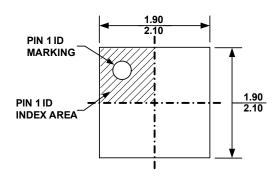


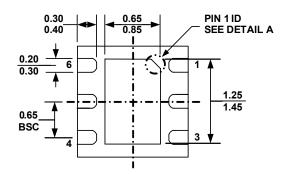
Figure 6: 5V Output Typical Application Circuit



PACKAGE INFORMATION

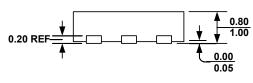
QFN-6 (2mmx2mm)



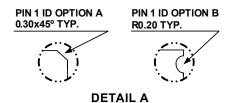


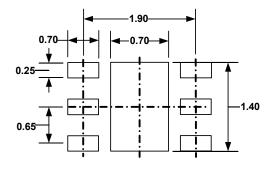
TOP VIEW

BOTTOM VIEW



SIDE VIEW





RECOMMENDED LAND PATTERN

NOTE:

- 1) ALL DIMENSIONS ARE IN MILLIMETERS 2) EXPOSED PADDLE SIZE DOES NOT INCLUDE MOLD FLASH
- 3) LEAD COPLANARITY SHALL BE0.10 MILLIMETER MAX
- 4) JEDEC REFERENCE IS MO-229, VARIATION VCCC.
- 5) DRAWING IS NOT TO SCALE

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